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CITATION:

Ikegami, Masanori ...[et al]. Double-slit Emittance Monitor for Pulsed Proton Beams.  
Bulletin of the Institute for Chemical Research, Kyoto University 1995, 73(1): 1-10

ISSUE DATE:

1995-03-31

URL:

<http://hdl.handle.net/2433/77603>

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## Double-slit Emittance Monitor for Pulsed Proton Beams

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*Received January 20, 1995*

At ICR, Kyoto University, an emittance monitor for beams from the DTL has been developed and installed in the beam line. With beam current of 0.7 mA at the entrance of the DTL, typical rms emittance for x direction is measured to be  $4.8 \pi \text{mm} \cdot \text{mrad}$ . Changing the RF power fed to the DTL from 250 kW to 300 kW, the rms emittance is measured to be varied from  $4.8 \pi \text{mm} \cdot \text{mrad}$  to  $9.0 \pi \text{mm} \cdot \text{mrad}$ . The variation of emittance in a macro pulse is also measured and the rms emittance is found to be varied from  $4.8 \pi \text{mm} \cdot \text{mrad}$  to  $5.9 \pi \text{mm} \cdot \text{mrad}$ .

KEY WORDS: Beam emittance/ Pulsed beam/ Two-slit method

### 1. INTRODUCTION

At ICR (Institute for Chemical Research), Kyoto University, a 7 MeV proton linac which consists of an RFQ linac and a DTL has been under improvement to get higher beam current<sup>1)</sup>. In these days, acceleration of intense charged particle beams in a linear accelerator is of considerable interest for variety of application. Beam evolution in a high current linac is strongly affected by the nonlinear space charge forces and resulting emittance growth limits the quality of the output beams. From the point of view, we have developed an emittance monitor for the beams from the DTL and installed it in the beam line.

Dependence of emittance on the RF power fed to the DTL is measured. Beam distribution in phase space is found to be affected by the strength of the RF power fed to the DTL. The variation of emittance within a macro pulse is also measured and the beam distribution in phase space is found to have some variation within a macro pulse.

Because the measured results were presented at a previous paper<sup>2)</sup>, in this paper the emphasis is put on the detailed hardware structure of the emittance monitor which could not be presented at the previous paper for lack of space.

### 2. EMITTANCE MONITOR

The emittance monitor is based on two-slit method and modified to measure the emittance of the pulsed beam with duration and repetition rate of  $50 \mu\text{sec}$  and 18~180 Hz, respectively<sup>2-4)</sup>. A schematic diagram of the emittance monitor is shown in Fig. 1 and its main specifications are

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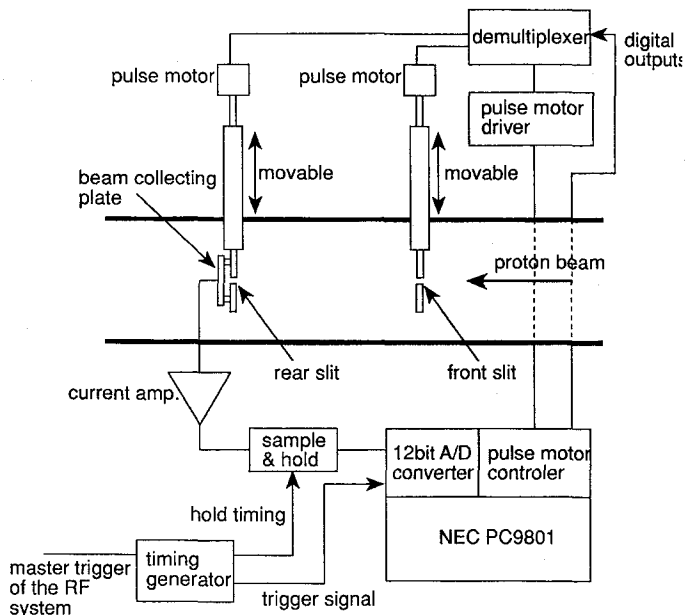


Fig. 1. A schematic diagram of the emittance monitor.

Table 1. Main specifications of the emittance monitor.

Slit width	0.2 mm
Slit thickness	5 mm
Distance between front and rear slit	1 m
Resolution of position	0.2 mm
Resolution of inclination angle	0.2 mrad
Displacement of slit	-20 mm ~ +80 mm (in steps of 0.2 mm)
Measurement range	$\pm 10 \text{ mm} \times \pm 10 \text{ mrad}$
Measurement time	5 min (1 mm $\times$ 1 mrad interval)

shown in Table 1. The emittance monitor has front and rear slit and the beam passed through the two slits is collected by a beam collecting plate. This current is amplified by a current amplifier, held by an S/H (sample and hold) circuit and led to a 12 bit A/D converter. The measurements of the beam current are synchronized to the RF timing signal applied to the DTL with use of the S/H circuit. Changing the hold timing, we can measure the variation of emittance within a macro pulse.

Two sets of the slits is installed in a vacuum chamber for  $x$  and  $y$  directions. The front slits for  $x$  and  $y$  directions are installed at positions of 1,475 mm and 1,325 mm downstream from the exit of the DTL, respectively. The overview of the emittance monitor is shown in Photo 1.

### 2.1 Slit and slit actuating system

The distance between the front and rear slit is 1 m and the widths and thicknesses of the collimating slits are 0.2 mm and 5 mm, respectively. The geometry of the front and rear slit are shown in Fig. 2. These slits are made of copper and water-cooled. These slits can be actuated

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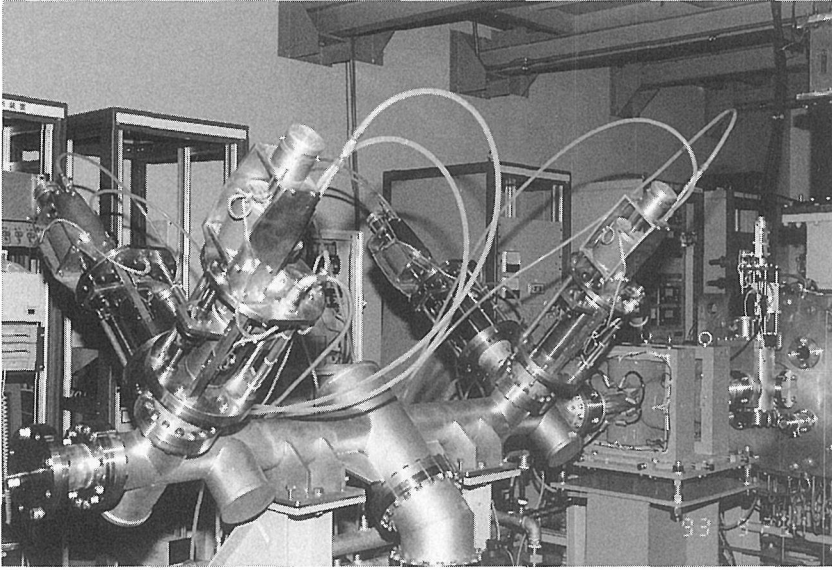


Photo 1. The overview of the emittance monitor.

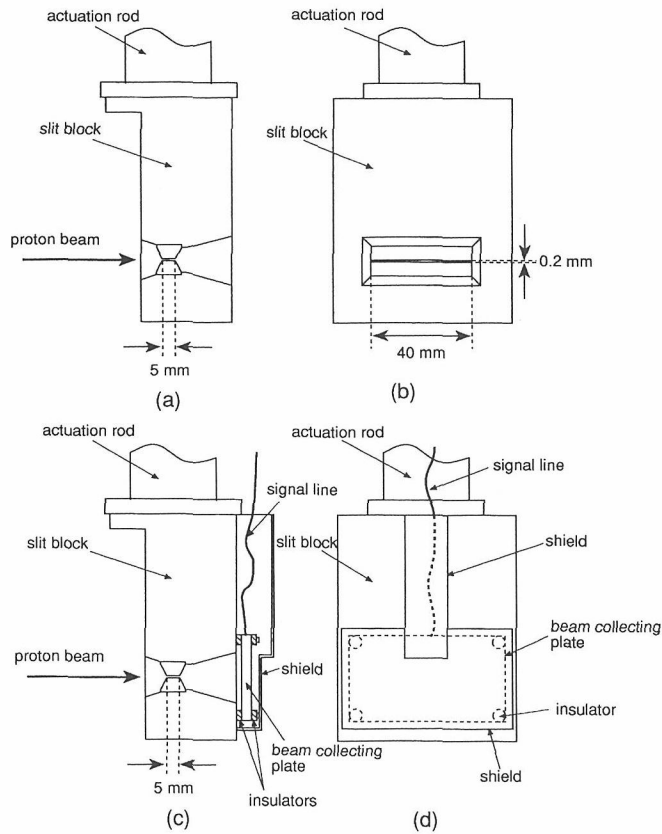


Fig. 2. The geometry of front and rear slit. (a) front slit, side view, (b) front slit, rear view, (c) rear slit, side view, (d) rear slit, rear view.

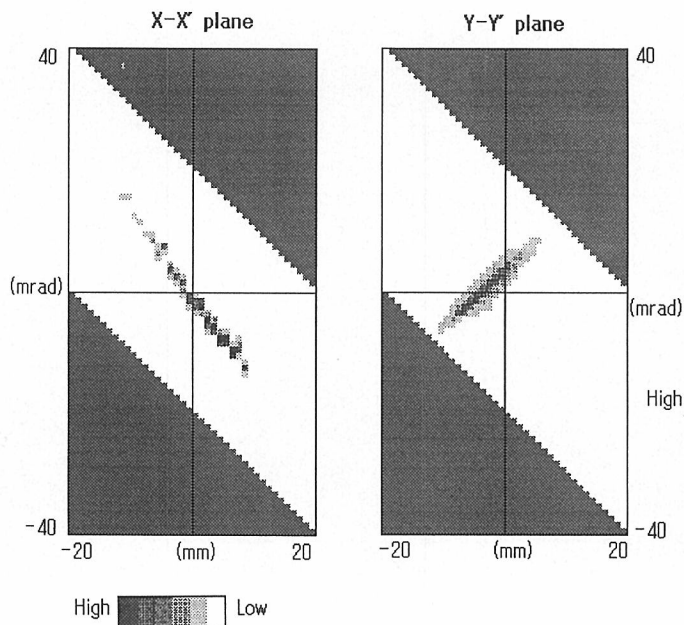


Fig. 3. An example of the contour plot. Scanning all the measurable region. White area represents the measurable region.

in steps of 0.2 mm with use of pulse motors and their total displacement from the design orbit can be varied from  $-20$  mm to  $+80$  mm. In this geometry, the maximum displacement and inclination angle of the beam to pass through the two slits is 20 mm and 40 mrad, respectively. The maximum speed of the slit actuation is 8 mm/sec. In the case of scanning the region of  $\pm 10$  mm  $\times$   $\pm 10$  mrad with 1 mm  $\times$  1 mrad intervals, it takes about 5 minutes to measure the contour plot of the beam intensity in phase space and calculate rms emittance for both planes. In the case of scanning all the measurable region, it takes about 12 minutes. An example of the contour plot is shown in Fig. 3.

We have a demultiplexer to drive 4 motors with one pulse motor driver. The demultiplexer consists of 9 relays and they are controlled by the digital outputs of the pulse motor controller. Four of the relays are used to control the electromagnetic breaks of the pulse motors and the others are used to switch the output of the pulse motor driver. The slit position can be measured by potentiometers with an accuracy of  $\pm 0.1$  mm.

## 2.2 Current detection system

The beam passed through the front and rear slit is collected by the beam collecting plate. The beam collecting plate is made of copper and shielded to suppress the noises possibly caused by secondary electrons (see Fig. 2 (c) and (d)). The shield is made of copper. The actuation rod, the slit block and the shield are electrically contacted to the ground and no suppressing voltage is applied.

The current collected by the beam collecting plate is amplified by the current amplifier. The circuit diagram of this amplifier is shown in Fig. 4. The current amplifier outputs a signal which is proportional to the input current and the inverting amplifier amplify the signal. The S/H circuit hold the output signal of the inverting amplifier. This type of circuit is adopted for

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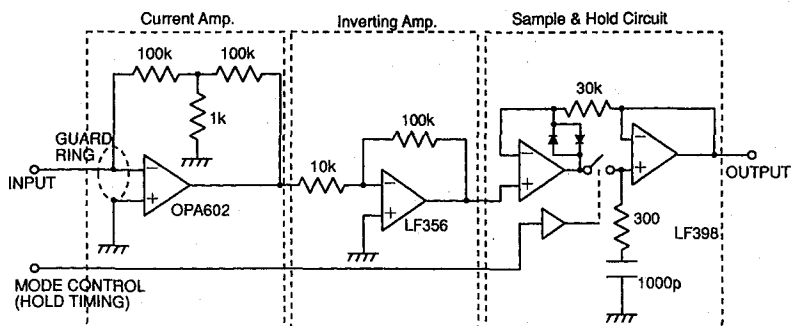
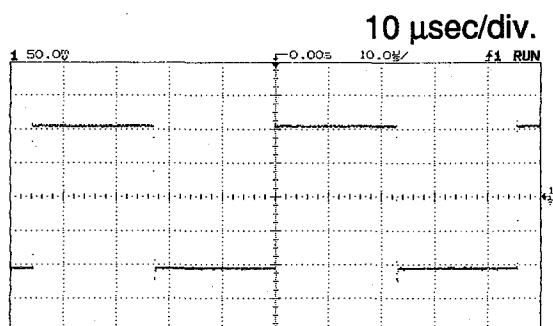
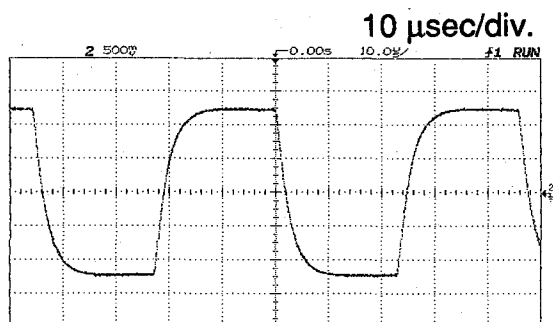


Fig. 4. The circuit diagram of the current amplifier.



(a)



(b)

Fig. 5. A typical output of the current amplifier: (a) input signal (peak current is 10 nA), (b) output signal

the current amplifier to shorten the time constant. Typical output signal and measured linearity of the current amplifier is shown in Fig. 5 and 6, respectively. With the beam current of 3~300 nA, the nonlinearity of the current amplifier is found to be negligible. The rise time of the amplifier is measured to be  $\sim 6 \mu\text{sec}$ .

The output of the current amplifier is held by the S/H circuit and led to the A/D converter. The hold timing of the S/H circuit and the trigger signal of the A/D converter is fed by the timing generator.

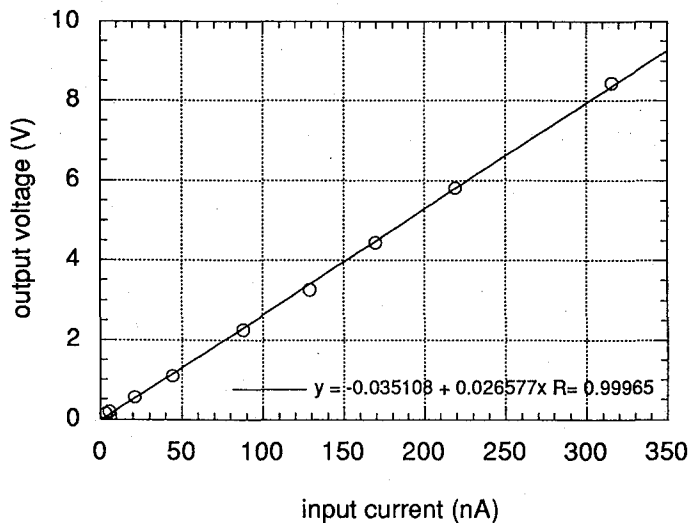


Fig. 6. The linearity of the current amplifier.

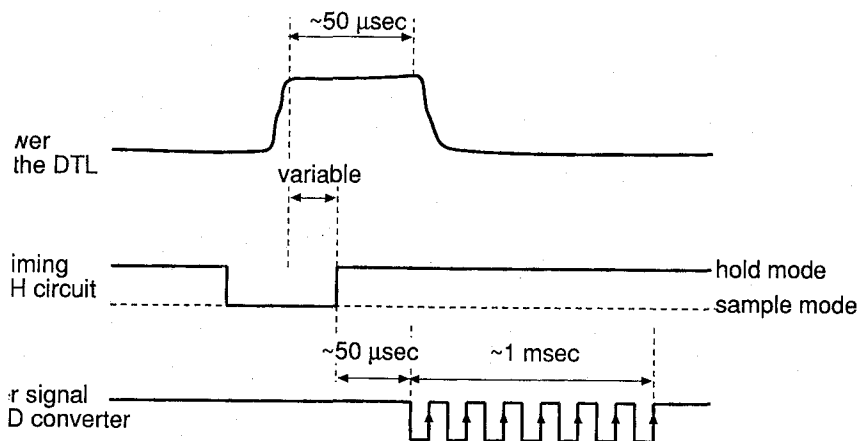


Fig. 7. The timing chart.

To care for noises, we can divide them to two parts. The one is the noises which are caused before the signal is held by the S/H circuit, and the other is the noises which are caused after the signal is held by the S/H circuit. The latter includes the quantization noise of the A/D converter. To smooth the latter noises, the A/D converter is triggered 7 times for every macro pulse. The timing chart is shown in Fig. 7. To smooth the former noises, we measure the beam current for 3 macro pulses and the average of them is adopted as the measured beam current for a measuring point in phase space.

### 3. DATA ANALYSIS

#### 3.1 Measurement of the offset voltage

To evaluate the offset voltage of the current amplifier, the output of the A/D converter is read

200 times for smoothing before each measurement. The offset voltage is measured with an accuracy of  $\pm 5$  mV. The accuracy of the measured emittance is mainly limited by the accuracy of the measurement of the offset voltage of the current amplifire. The accuracy of the measured emittance depends on the beam current at the exit of the DTL. With beam current of 0.6 mA at the exit of the DTL, we can measure the emittance with an accuracy of  $\pm 1 \pi$  mm·mrad.

### 3.2 Correction for the slit thickness

For beams which have large inclination angle, the effective slit width is smaller than the actual slit width because of the slit thickness. So, the measured beam current is corrected by the relation,

$$I_{cor} = \frac{\frac{d}{l}}{\frac{d}{l} - |x'|} I_{mes}$$

where  $I_{mes}$ ,  $I_{cor}$ ,  $x'$ ,  $d$  and  $l$  denote measured current, corrected current, inclination angle, slit width and slit thickness, respectively.

### 3.3 Calculation of the rms emittance

There are several definitions for rms emittance of ion beams. We adopt the following definition,

$$\epsilon_{rms}^2 = \langle (x - \langle x \rangle)^2 \rangle \langle (x' - \langle x' \rangle)^2 \rangle - \langle (x - \langle x \rangle)(x' - \langle x' \rangle) \rangle^2$$

$$\langle A(x, x') \rangle \equiv \frac{\iint A(x, x') g(x, x') dx dx'}{\iint g(x, x') dx dx'}$$

where  $g(x, x')$  denotes the continous phase space distribution function of the beam and  $A(x, x')$  denotes an arbitrary function of  $x$  and  $x'$ . It should be noted that some authors introduce the factor of 4 so that the rms emittance gives correct emittance when applied to an ideal distribution, a uniformly filled ellipse. It should also be noted that the integration in the above equation should be replaced by summation because the experimental measurement of  $g(x, x')$  is done discretely.

## 4. EXPERIMENTAL SETUP

At ICR proton linac, output beam quality is strongly affected by the LEBT (low energy beam transport) parameters such as applied voltage on the electric quadrupoles and the Einzel lens. During the experiments, the LEBT parameters are fixed to a set of values which enables us to accelerate relatively higher current of beam.

In the measurement, the dependence of emittance on the RF power fed to the DTL and the hold timing are measured<sup>2)</sup>. While measuring the dependence of emittance on the RF power fed to the DTL, hold timing and beam current at the entrance of the DTL is fixed to  $t_3$  (see Fig. 8) and 0.7 mA, respectively. While measuring the dependence of the hold timing, beam current at the entrance of the DTL and RF power fed to the DTL is fixed to 0.7 mA and 300 kW, respectively.

Between the emittance monitor and the exit of the DTL a newly fabricated steering magnet<sup>5)</sup> and a quadrupole doublet are installed, and these magnets are used to steer the beam and get



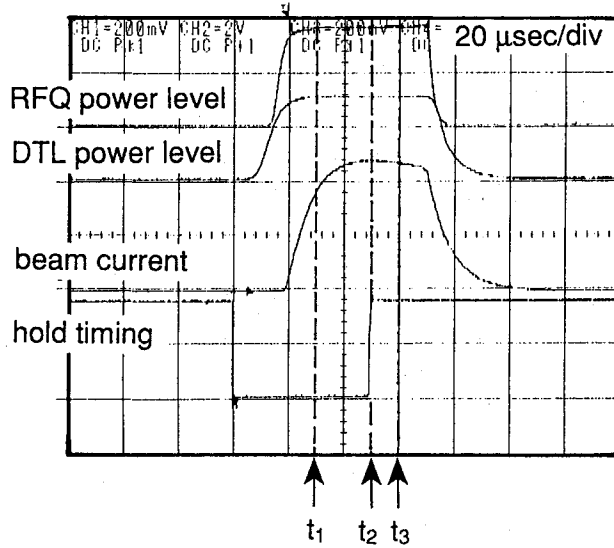


Fig. 8. The definition of the hold timing,  $t_1 \sim t_3$ .

moderate focusing at the emittance monitor.

The emittance is measured scanning all the measurable region with  $1\text{ mm} \times 1\text{ mrad}$  intervals.

## 5. MEASURED RESULTS

### 5.1 Dependence on the RF power

Changing the RF power fed to the DTL from 250 kW to 300 kW, the rms emittance is measured to be varied from  $4.8\pi\text{ mm}\cdot\text{mrad}$  to  $9.0\pi\text{ mm}\cdot\text{mrad}$ . Dependence of emittance on the RF power fed to the DTL is shown in Fig. 9. The figure tells us that emittance increases as the RF power is lowered, and at the same time, position of the beam centroid is shifted in phase space. These shifts can be explained by a displacement of the beam centroid at the entrance of the DTL. Beams having a displacement at the entrance of the DTL are evolved oscillating around the design orbit under an influence of RF defocusing forces. When the RF power fed to the DTL is lowered, the RF defocusing force decreases and it causes shift of the betatron tune.

Simulations performed with use of the PARMILA code show that about 1 mm displacement of the beam centroid at the entrance of the DTL will cause dependence of beam distribution in the phase space on the RF power with approximately the same amplitude which is observed at the measurement, but 1 mm displacement is not consistent with the accuracy of the alignment. So, other possibilities such as alignment errors of PMQs (permanent quadrupole magnets) should be checked.

At the measurement, the transmission ratio of the DTL is measured to be  $\sim 70\%$ . So, the lost part of the beams may be responsible for the change of the measured emittance.

### 5.2 Variation in the macro pulse

At ICR proton linac, beams have macro pulse length of  $50\mu\text{sec}$  and the RF power fed to the DTL vary within the pulse. Considering measured results of the previous subsection, this

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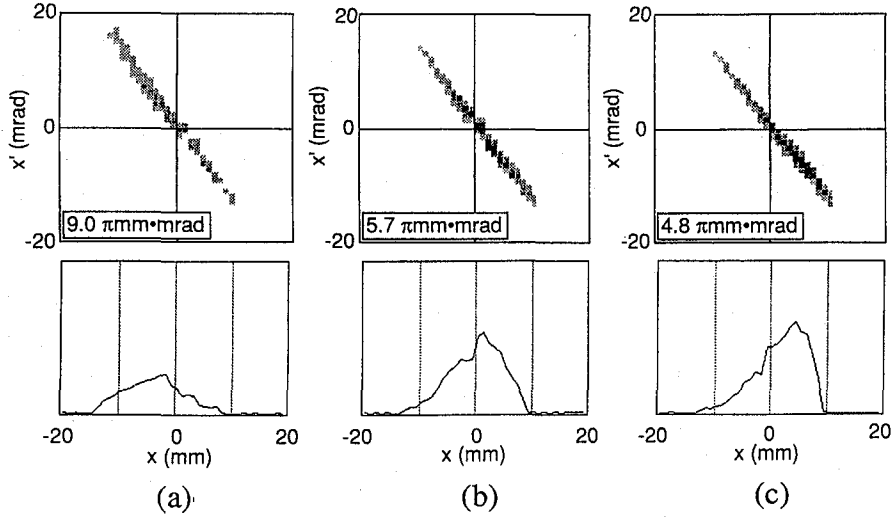


Fig. 9. The dependence of the emittance on the RF power fed to the DTL,  $P_{in}$ . (a)  $P_{in}=250 \text{ kW}$ , (b)  $P_{in}=275 \text{ kW}$ , (c)  $P_{in}=300 \text{ kW}$  (design value). The rms emittance is shown in below left and the projected distribution in real space is shown in bottom.

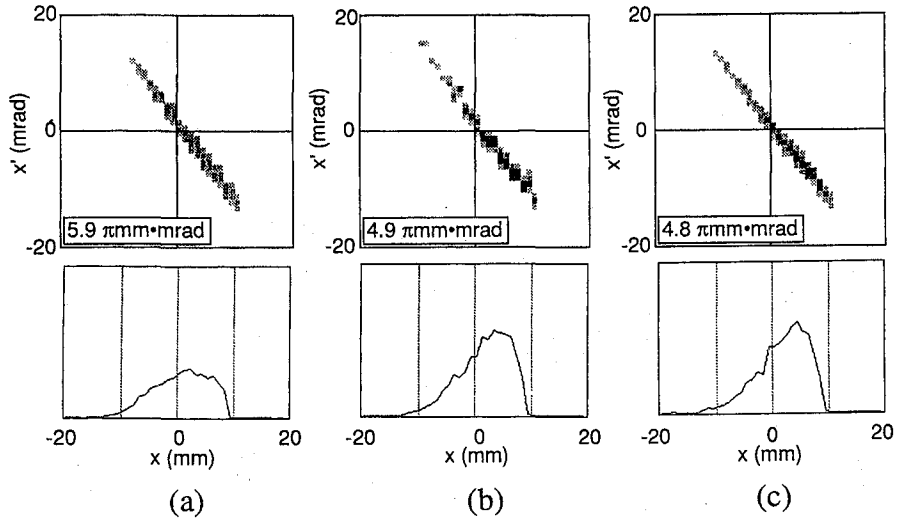


Fig. 10. The dependence of the emittance on the hold timing,  $t_h$ . (a)  $t_h=t_1$ , (b)  $t_h=t_2$ , (c)  $t_h=t_3$ . The rms emittance is shown in below left and the projected distribution in real space is shown in bottom.

variation is expected to affect the beam distribution in phase space. From that point of view, we examine the variation of the emittance within the pulse changing the hold timing of the S/H circuit. The rms emittance is measured to be varied from  $4.8 \pi \text{ mm} \cdot \text{mrad}$  to  $5.9 \pi \text{ mm} \cdot \text{mrad}$  in a macro pulse. Measured dependence of emittance on hold timing is shown in Fig. 10. The definition of the hold timing,  $t_1 \sim t_3$  is shown in Fig. 8. In Fig. 10, we can see a variation of emittance within the macro pulse. Though no apparent variation of the position of beam

centroid is found, the variation of emittance is relatively large, So, it seems that the variation is difficult to be explained only by the variation of the RF power within the pulse. We should examine the dependence of the variation on the RF phase control and other RF parameters.

Because the current amplifier has a time constant of  $6\mu\text{sec}$ , measured emittance should be considered to be an averaged value over  $\pm 3\mu\text{sec}$  around the hold timing. So, it is difficult to examine the detailed structure with the current amplifier.

## 6. SUMMARY

We developed an emittance monitor for beams from the DTL. Dependence of the emittance on the RF power fed to the DTL and the hold timing of the S/H circuit is measured. Measured results can be summarized as follows:

1) Beam distribution in phase space is found to have a dependence on the RF power fed to the DTL. Changing the RF power fed to the DTL from 250 kW to 300 kW, the rms emittance is measured to be varied from  $4.8\pi\text{mm}\cdot\text{mrad}$  to  $9.0\pi\text{mm}\cdot\text{mrad}$ .

2) It is found that there is a variation of the beam distribution in phase space within a macro pulse. The rms emittance is measured to be varied from  $4.8\pi\text{mm}\cdot\text{mrad}$  to  $5.9\pi\text{mm}\cdot\text{mrad}$  in a macro pulse.

In future experiments, we plan to measure the emittance growth in the linac quantitatively. For this purpose, an emittance monitor to measure the emittance at the exit of the ion source is now under development.

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